

A glimpse at Sampled-Data systems

Nothing new, just some elements from literature

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FEANICESES 2019 Workshop

Idea behind the research team COVNI:



computer theory

+



automatic control

=



something interesting for
computer-controlled systems

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[2] Chen and Francis. *Optimal sampled-data control systems*. 1995

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How do we assess dynamical performances of an implemented control-law?

- ▶ give some insights concerning what "performance" may refer to for control people
- ▶ introduce some concepts from sampled-data systems theory
- ▶ talk restricted to Linear Time Invariant (LTI) models

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References:

- ▶ a comprehensive introduction to control engineering can be found in the first chapters (1 → 3) of [1]
- ▶ what will be presented on sampled-data systems can be found in [2]

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Outline

Performances of a controlled system

Quantifying discretization error

Continuous lifting and measure of performance for SD systems

Conclusion

What is control engineering all about? I

In a nutshell

dynamical system + actuators/sensors + control objectives = *control problem* (1)
tracking, disturbance rejection, etc.

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And in practice...?

Different philosophies: pole placement, loop-shaping, convex synthesis...

↪ *Here: focus on the approach relying on the specification of a performance function to be optimized ~ optimal feedback control*

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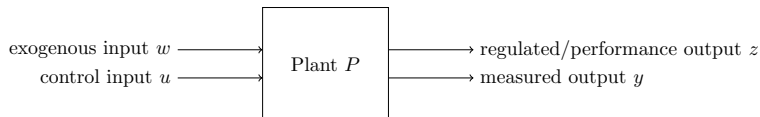
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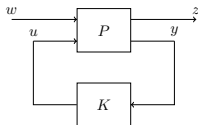
Modelling of the control problem = derive the generalized plant P :

- ▶ models for the dynamical system + actuators/sensors → this sets u and y
- ▶ identify reference signals, possible disturbances or noise → w
- ▶ select signals of interest for the control objective → z

What is control engineering all about? II

Standard form

The loop is closed with some controller K ,

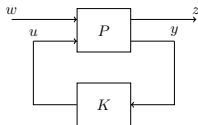


- ▶ most of the control architectures can be represented as this standard form,
- ▶ K can have different structures: a gain, PI, PID, etc.
- ▶ the closed-loop is the transfer from w to z , denoted $T_{w \rightarrow z}(P, K)$,
- ▶ in general, one wants K to stabilize (internally) the plant P .

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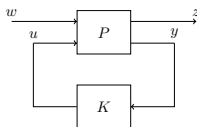
Find K that stabilizes P internally and minimizes $\|T_{w \rightarrow z}\|$

For fixed plant P , performance of the closed-loop = value of $\|T_{w \rightarrow z}\|$

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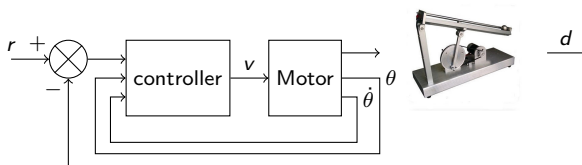
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Linear-Time Invariant case

- ▶ practical interpretation of $T_{w \rightarrow z}$ convenient in the frequency-domain (transfer function)
- ▶ useful systems norms: $\mathcal{H}_\infty, \mathcal{H}_2$

What is control engineering all about? III

A tracking example: ball on the beam

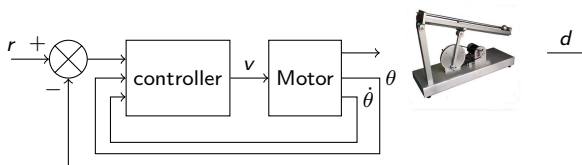


Physical modelling

- ▶ dynamical system: beam + ball,
- ▶ actuator: motor e.g. controlled by a voltage $v \rightarrow u$
- ▶ sensed signals: the distance d of the ball to the center of the beam, the angular position θ and velocity $\dot{\theta}$ of the shaft of the motor $\rightarrow y = [r - d, \theta, \dot{\theta}]^T$

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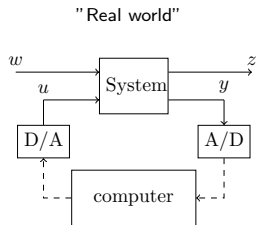
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Signals for the control problem:

- ▶ tracking objective $\rightarrow z = r - d$
- ▶ only exogenous input $w = r$

The *performance channel* $T_{w \rightarrow z}$ describes how fast the system will follow the reference. It can be filtered, e.g. with a low-pass filter.

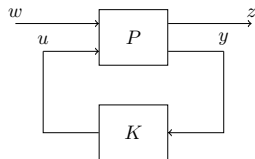
Fifty shades of models



► need a model

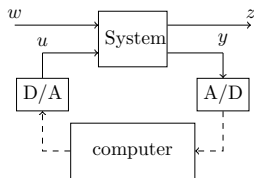
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Continuous-time/analog
modelling



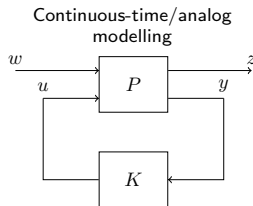
- ▶ idealized (LTI)
- ▶ lots of techniques for design, analysis, etc.

"Real world"

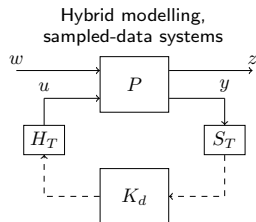


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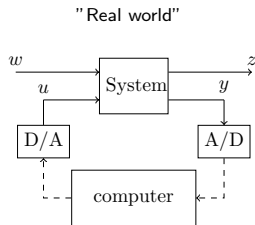
Fifty shades of models



- ▶ idealized (LTI)
- ▶ lots of techniques for design, analysis, etc.



- ▶ less idealized (still...)
- ▶ more involved to manipulate (not LTI!)



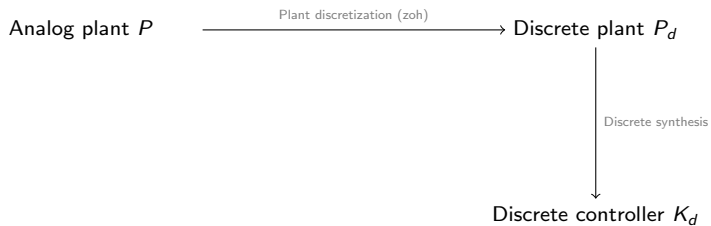
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Optimize then discretize...?

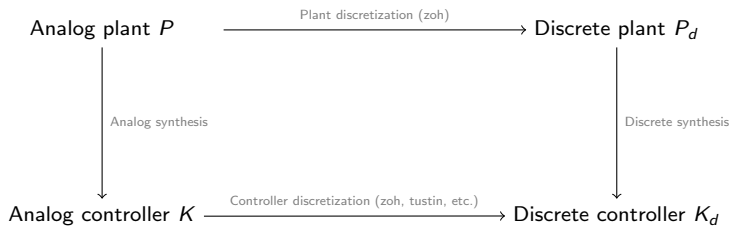
Analog plant P

Discrete controller K_d

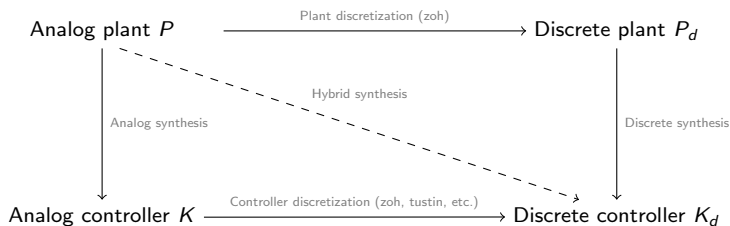
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- ▶ indirect methods lose optimality
- ▶ intersample behaviour is discarded
- ▶ decreasing T requires higher computational power and accuracy
↳ with shift operator, $\lim_{T \rightarrow 0} A_d = I$ and $\lim_{T \rightarrow 0} B_d = 0$

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Ideal sampler & holder

Reminder: LTI model G , i/o relationship in frequency domain

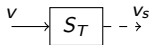
$$\text{analog: } Y(j\omega) = G_a(j\omega)U(j\omega) \quad \text{discrete: } Y(e^{j\omega}) = G_d(e^{j\omega})U(e^{j\omega})$$

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Sampler S_T



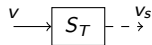
- ▶ time domain: $v_s[k] = v(kT)$
- ▶ modelled by an impulse-train $\sum_k \delta(t - kT) = T$ -periodic function
- ▶ frequency domain: $V_s(e^{-j\omega T}) = \frac{1}{T} \sum_{k \in \mathbb{Z}} V(j\omega + jk\omega_s)$ frequency aliasing
- ▶ /!\ not bounded for all signals in \mathcal{L}_2 (bandlimited signals or anti-aliasing filter)

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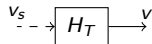
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Holder H_T



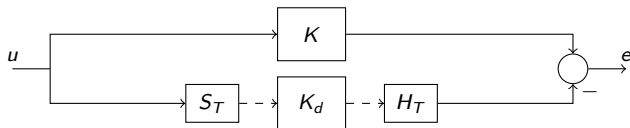
- ▶ time domain: $v(t) = v_s[k], kT \leq t < (k+1)T$
- ▶ modelled as a difference of delayed steps: $h_T(t) = \frac{1}{T} \mathbf{1}(t) - \frac{1}{T} \mathbf{1}(t - T)$
- ▶ frequency domain: $H_T(s) = \frac{1 - e^{-sT}}{sT}$ and $V(j\omega) = TH_T(j\omega)V_s(e^{-j\omega T})$

Discretization error

Let K be an analog LTI model and K_d its discretization: **Meaningful notion of error?**

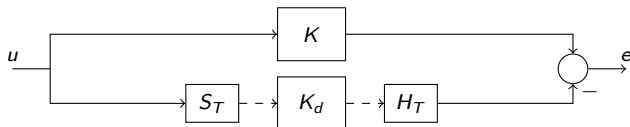
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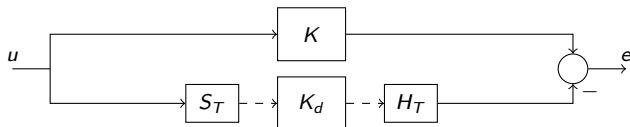


- ▶ the error system is time-varying (T -periodic) \rightarrow no transfer function
- ▶ $H_T S_T$ is not bounded for all signals in \mathcal{L}_2 \rightarrow need bandlimited signal u or anti-aliasing filter
- ▶ frequency-domain error:

$$E(j\omega) = K(j\omega)U(j\omega) - H_T(j\omega)K_d(e^{-j\omega T}) \sum_k U(j\omega + jk\omega_s) \quad (2)$$

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suppose $U(j\omega) = 0$ for $\omega > \omega_N$ (bandlimited),

$$E(j\omega) = \left(K(j\omega) - H_T(j\omega)K_d(e^{-j\omega T}) \right) U(j\omega) \quad (3)$$

\rightarrow error(ω) = $|K(j\omega) - H_T(j\omega)K_d(e^{-j\omega T})|$

Discretization error: example

$$K(s) = \frac{1}{1 + 0.7s/10 + s^2/100} \quad (4)$$

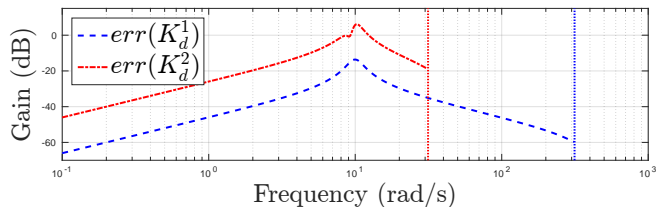
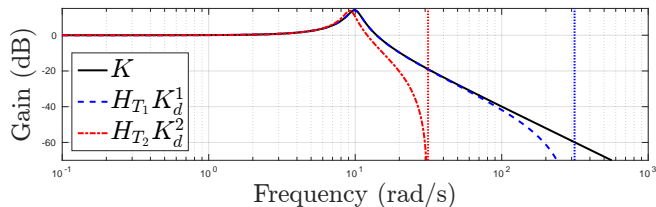
- ▶ $T_1 = 0.01 \rightarrow K_d^1$
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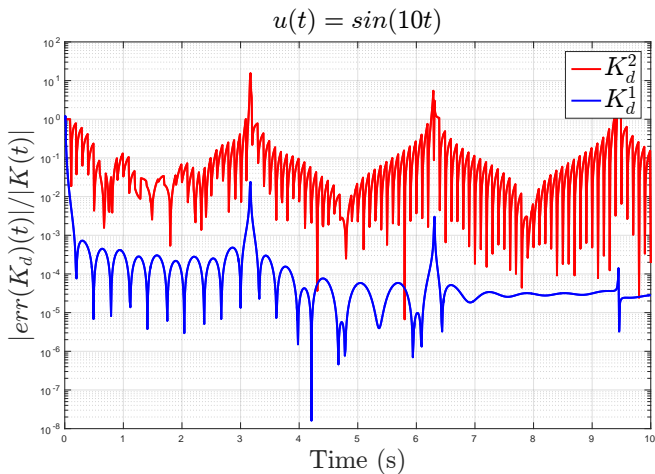


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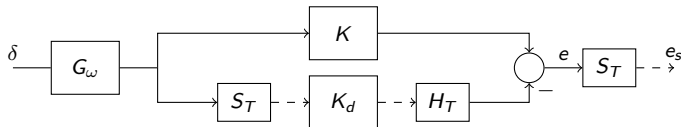
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Optimal discretization?

- ▶ frequency-domain error not usual system norm
- ▶ possible to look for the best discretization scheme w.r.t. the sampling instants and w.r.t. a predefined class of signals
 - ↳ section 4.6 in [2]
 - ↳ usual discrete norm can be used to measure e_s
 - ↳ e.g. step invariant discretization (ZOH) is the solution for steps



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Frequency-domain considerations are central in some design methods...but SD have no transfer function

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Continuous lifting

Consider a signal $u(t)$ such that $\int_{-\infty}^{\infty} u(t)^2 dt < \infty$ (i.e. $\in \mathcal{L}_2(\mathbb{R})$)

1. divide u in chunks of length T (sampling time): $\dots, [-T, 0), [0, T), [T, 2T), \dots$

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Lifting operator:

$$\begin{aligned} L : \mathcal{L}_2(\mathbb{R}) &\rightarrow l_2(\mathbb{Z}, \mathcal{K}) \\ u(t) &\rightarrow \underline{u} \end{aligned} \tag{5}$$

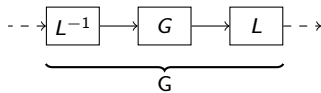
- it turns out L is an isomorphism
↔ working with u or \underline{u} is 'equivalent'

Lifting systems I

$$G : \begin{cases} \dot{x}(t) &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) + Du(t) \end{cases} \quad \text{such that } y = Gu \quad (6)$$

Lifted system \underline{G} obtained by lifting input/output

- ▶ $\underline{u} = Lu$
- ▶ $\underline{y} = Ly$
- ▶ thus $\underline{y} = \underline{G}\underline{u}$ with $\underline{G} = LGL^{-1}$

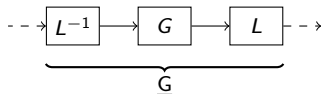


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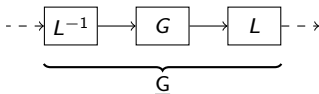
$$\underline{G} : \begin{cases} \xi_{k+1} &= \underline{A}\xi_k + \underline{B}u_k \\ y_k &= \underline{C}\xi_k + \underline{D}u_k \end{cases} \quad \text{with } \xi_k = x(kT) \quad (7)$$

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where the transformations are

$$\begin{aligned} \underline{A} : & \mathcal{E} \rightarrow \mathcal{E}, & \underline{A}\xi &= e^{TA}\xi \\ \underline{B} : & \mathcal{K} \rightarrow \mathcal{E}, & \underline{B}u &= \int_0^T e^{(T-\tau)A}Bu(\tau)d\tau \\ \underline{C} : & \mathcal{E} \rightarrow \mathcal{K}, & (\underline{C}\xi)(t) &= Ce^{tA}\xi \\ \underline{D} : & \mathcal{K} \rightarrow \mathcal{K}, & (\underline{D}u)(t) &= Du(t) + \int_0^t Ce^{(t-\tau)A}Bu(\tau)d\tau \end{aligned} \quad (8)$$

Lifting systems II

Interest?

- ▶ D_T and D_T^* : time delay and time advance of T
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$$LD_T^* = U^*L \quad \text{and} \quad L^{-1}U = D_T L^{-1} \quad (9)$$

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G is T -periodic $\Leftrightarrow \underline{G}$ is time invariant

Proof (\Rightarrow)

$$\begin{aligned} U^* \underline{G} U &= U^* L G L^{-1} U \\ &= LD_T^* G D_T L^{-1} \\ &= L G L^{-1} \\ &\triangleq \underline{G} \end{aligned} \quad (10)$$

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$$LD_T^* = U^*L \quad \text{and} \quad L^{-1}U = D_T L^{-1} \quad (9)$$

G is T -periodic $\Leftrightarrow \underline{G}$ is time invariant

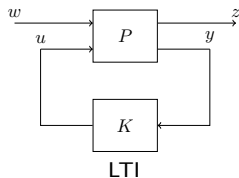
Proof (\Rightarrow)

$$\begin{aligned} U^* \underline{G} U &= U^* L G L^{-1} U \\ &= L D_T^* G D_T L^{-1} \\ &= L G L^{-1} \\ &\triangleq \underline{G} \end{aligned} \quad (10)$$

A SD system can be lifted into a time-invariant discrete-time system with infinite-dimensional input and output spaces

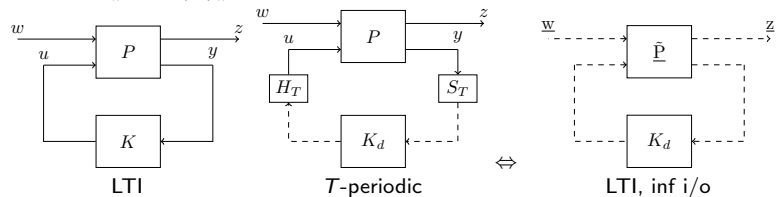
Measure of performance of closed-loop SD systems I

Reminder: $\|T_{w \rightarrow z}(K)\|$ sums up the performances of the closed-loop



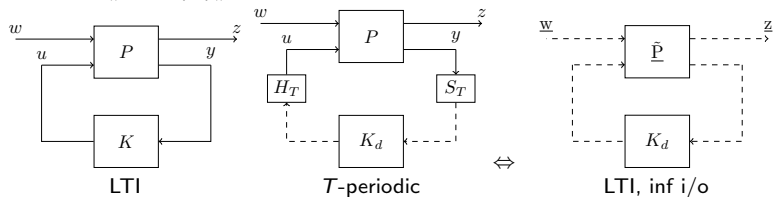
Measure of performance of closed-loop SD systems I

Reminder: $\|T_{w \rightarrow z}(K)\|$ sums up the performances of the closed-loop



Measure of performance of closed-loop SD systems I

Reminder: $\|T_{w \rightarrow z}(K)\|$ sums up the performances of the closed-loop



In chap. 12-13 of [2]:

- ▶ similar norms (2 and ∞) can be defined for $T_{w \rightarrow z}(K_d)$ through $T_{\underline{w} \rightarrow \underline{z}}(K_d)$
- ▶ ...leading to some optimal synthesis framework
- ▶ ...and analysis
- ▶ but more difficult due to infinite dimensionality of input/output spaces

Measure of performance of closed-loop SD systems II

Some interesting applications in [2]

Consider two discretizations $K_d^1(T)$ and $K_d^2(T)$ of some controller K

- ▶ compare their performances against sampling rate T
↪ example 13.7.3
- ▶ determine T_{max} such that some prescribed performance bound is fulfilled
↪ example 13.8.1
- ▶ perform optimal synthesis in SD framework
↪ example 13.8.1

Measure of performance of closed-loop SD systems II

Some interesting applications in [2]

Consider two discretizations $K_d^1(T)$ and $K_d^2(T)$ of some controller K

- ▶ compare their performances against sampling rate T
↪ example 13.7.3
- ▶ determine T_{max} such that some prescribed performance bound is fulfilled
↪ example 13.8.1
- ▶ perform optimal synthesis in SD framework
↪ example 13.8.1

Other?

- ▶ given P, K , for fixed T , find K_d that minimizes the loss of performance w.r.t. analog design

$$\|T_{\underline{w} \rightarrow \underline{z}}(K) - T_{\underline{w} \rightarrow \underline{z}}(K_d)\| \quad (11)$$

↪ everything must be considered in the lifted domain

Outline

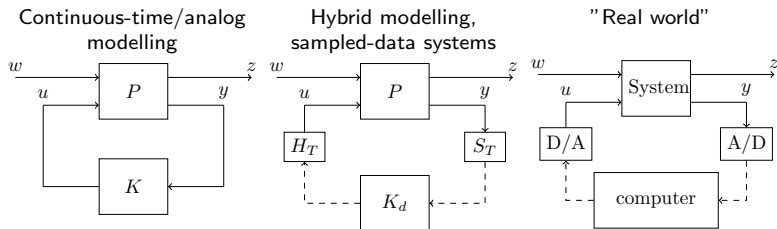
Performances of a controlled system

Quantifying discretization error

Continuous lifting and measure of performance for SD systems

Conclusion

Conclusion



- ▶ Sampled-Data formalism suited for computer-controlled systems
- ▶ Lifting technique to extend LTI tools
- ▶ Lifting can also be used to handle multiple sampling rates, chap. 8 [2]
- ▶ infinite dimensionality...
- ▶ still not implemented controller (code)...





Tools: [3], [4]

[2] Chen and Francis. *Optimal sampled-data control systems*. 1995

[3] Fujioka, Hara, and Yamamoto. "Sampled-data control toolbox: object-oriented software for sampled-data feedback control systems". 2004

[4] Polyakov, Rosenwasser, and Lampe. "DIRECTSD 3.0 toolbox for MATLAB: Further progress in polynomial design of sampled-data systems". 2006

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-  T. Chen and B.A. Francis. *Optimal sampled-data control systems*. Springer Science & Business Media, 1995.
-  H. Fujioka, S. Hara, and Y. Yamamoto. “Sampled-data control toolbox: object-oriented software for sampled-data feedback control systems”. In: *International Conference on Robotics and Automation*. 2004, pp. 19–24.
-  K. Polyakov, E. Rosenwasser, and B. Lampe. “DIRECTSD 3.0 toolbox for MATLAB: Further progress in polynomial design of sampled-data systems”. In: *Conference on Computer Aided Control System Design*. 2006, pp. 1946–1951.